

INTERNATIONAL STANDARD

NORME INTERNATIONALE

Assessment methods of the human exposure to electric and magnetic fields from wireless power transfer systems – Models, instrumentation, measurement and computational methods and procedures (frequency range of 3 kHz to 30 MHz)

Méthodes d'évaluation de l'exposition humaine aux champs électriques et magnétiques produits par les systèmes de transfert de puissance sans fil – Modèles, instrumentation, méthodes et procédures de mesure et de calcul (Plage de fréquences comprise entre 3 kHz et 30 MHz)



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ASSESSMENT METHODS OF THE HUMAN EXPOSURE TO ELECTRIC AND MAGNETIC FIELDS FROM WIRELESS POWER TRANSFER SYSTEMS – MODELS, INSTRUMENTATION, MEASUREMENT AND COMPUTATIONAL METHODS AND PROCEDURES (FREQUENCY RANGE OF 3 kHz TO 30 MHz)

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This edition includes the following significant technical changes with respect to the previous edition:

- a) lower frequency bound changed from 1 kHz to 3 kHz;
- b) clarified contact currents as indirect effects in assessment procedures;
- c) in measurement methods applied the formulas of SAR and internal electric field;
- d) in computational assessment methods added specifications for averaging of current density and internal E-field;
- e) updated uncertainty of computational methods;
- f) introduced test reporting contents guidance.

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- revised.

INTRODUCTION

The wireless power transmission systems described in the scope of this document require particularly developed procedures and protocols for the assessment of human exposure. Such systems are increasingly being implemented in a wide range of applications at different frequency ranges from consumer electronics (e.g. mobile phones, tablet PCs) to automotive (electric vehicles). Human exposure to electric and magnetic fields is limited to avoid established adverse health effects, including electrostimulation of nervous tissues and thermal effects, as well as contact currents. A published ITU-R report (ITU-R SM.2303-3 [1]¹) on WPT systems specifies RF exposure assessment methodologies, yet no definitive assessment method was introduced. An exposure assessment method of WPT for EV charging systems was specified in IEC 61980-3:2022 [2]; however, there are currently no other detailed product standards related to WPT systems. Because WPT systems will continue to become ubiquitous in a multitude of applications in the future, IEC and IEEE established a joint working group to address WPT system assessment methods related to human exposures to electric, magnetic, and electromagnetic fields.

In this document, the basic methods to assess both direct and indirect effects of exposure to WPT systems, case studies, and relevant research are specified. These methods mainly focus on frequencies between 3 kHz and 30 MHz and consider both electrostimulation and thermal effects. Future editions will consider extended guidance for assessments of exposure from capacitive WPT systems.

¹ Numbers in square brackets refer to the Bibliography.

ASSESSMENT METHODS OF THE HUMAN EXPOSURE TO ELECTRIC AND MAGNETIC FIELDS FROM WIRELESS POWER TRANSFER SYSTEMS – MODELS, INSTRUMENTATION, MEASUREMENT AND COMPUTATIONAL METHODS AND PROCEDURES (FREQUENCY RANGE OF 3 kHz TO 30 MHz)

1 Scope

The objective of this document is to specify methods to assess human exposure to electromagnetic fields generated by stationary wireless power transfer (WPT) in terms of specific absorption rate (SAR), internal electric fields² or current density, and contact currents. The frequency range covered by this document is from 3 kHz to 30 MHz. This document focuses on exposures from inductive WPT systems and specifies:

- general compliance assessment procedures;
- measurement methods;
- computational assessment methods;
- assessment combining measurement and computational methods.

This document does not consider the immunity of cardiac implantable electrical devices to radiated disturbances from WPT systems.

2 Normative references

The following documents are referred to in the text in such a way that some or all of their content constitutes requirements of this document. For dated references, only the edition cited applies. For undated references, the latest edition of the referenced document (including any amendments) applies.

IEC 61786-1:2013, *Measurement of DC magnetic, AC magnetic and AC electric fields from 1 Hz to 100 kHz with regard to exposure of human beings – Part 1: Requirements for measuring instruments*

IEC 61786-1:2013/AMD1:2024

IEC 61786-2:2014, *Measurement of DC magnetic, AC magnetic and AC electric fields from 1 Hz to 100 kHz with regard to exposure of human beings – Part 2: Basic standard for measurements*

IEC/IEEE 62209-1528:2020, *Measurement procedure for the assessment of specific absorption rate of human exposure to radio frequency fields from hand-held and body-mounted wireless communication devices – Part 1528: Human models, instrumentation, and procedures (Frequency range of 4 MHz to 10 GHz)*

IEC/IEEE 62704-1:2017, *Determining the peak spatial-average specific absorption rate (SAR) in the human body from wireless communications devices, 30 MHz to 6 GHz – Part 1: General requirements for using the finite difference time-domain (FDTD) method for SAR calculations*

² Internal electric field is associated with exposure assessments of nerve stimulation effects; further information is available in e.g. [5].

IEC/IEEE 62704-4:2020, *Determining the peak spatial-average specific absorption rate (SAR) in the human body from wireless communications devices, 30 MHz to 6 GHz – Part 4: General requirements for using the finite element method for SAR calculations*

3 Terms and definitions

For the purposes of this document, the following terms and definitions apply.

ISO, IEC, and IEEE maintain terminology databases for use in standardization at the following addresses:

- IEC Electropedia: available at <https://www.electropedia.org/>
- ISO Online browsing platform: available at <https://www.iso.org/obp>
- IEEE Dictionary Online: available at <http://dictionary.ieee.org>

3.1

exposure

<of a body> situation that occurs wherever a person is subjected to electric, magnetic, or electromagnetic fields

3.2

basic restriction

BR

maximum permissible field level induced in the body of a person exposed to electric, magnetic or electromagnetic fields below which established adverse health effects are assumed not to occur

Note 1 to entry: Examples of basic restrictions can be found in International Commission on Non-Ionizing Radiation Protection (ICNIRP) Guidelines [3], [4], [5] and in Annex II of the European Union (EU) Council Recommendation 1999/519/EC [6].

Note 2 to entry: Documents issued by the IEEE ICES TC 95, such as [7] and [8], refer to the basic restrictions as dosimetric reference limits (DRL).

Note 3 to entry: European Union (EU) Directive 2013/35/EU [9] refers to basic restrictions as "exposure limit values" (ELVs), i.e. "values established on the basis of biophysical and biological considerations, in particular on the basis of scientifically well-established short-term and acute direct effects, i.e. thermal effects and electrical stimulation of tissues."

3.3

reference level

maximum permissible level of the incident electric, magnetic, or electromagnetic fields below which the basic restrictions are assumed not to be exceeded

Note 1 to entry: Examples of reference levels can be found in ICNIRP Guidelines [3], [4], [5] and in Annex II of the European Union (EU) Council Recommendation 1999/519/EC [6].

Note 2 to entry: Documents issued by the IEEE ICES TC 95, such as [7] and [8], refer to the reference levels as exposure reference levels (ERL).

Note 3 to entry: European Union (EU) Directive 2013/35/EU [9] refers to reference levels as "action levels (ALs)," i.e. "operational levels established for the purpose of simplifying the process of demonstrating the compliance with relevant ELVs or, where appropriate, to take relevant protection or prevention measures specified in this Directive."

3.4

direct effect

biological effect resulting from exposure of the body to electric, magnetic, or electromagnetic fields without interaction with a conducting object

3.5**indirect effect**

effect caused by the presence of an object or a substance in an electromagnetic field, which might present a safety or health hazard

3.6**incident field**

field that corresponds to the field that would have been present in the absence of an exposed person or phantom

Note 1 to entry: In some documents, the incident field is called an unperturbed field or environmental field.

3.7**coupling factor**

factor to conservatively convert an incident field level into an induced field level

Note 1 to entry: Generally, coupling factors are determined by evaluating the correlation between incident and induced fields for a large number of exposure configurations.

3.8**coverage factor**

factor that is applied in a physical relation to conservatively compensate for effects that are not considered in the formulation of the relation itself and would lead to an overly complicated formulation if evaluated accurately

3.9**electric field strength**

vector field quantity E which exerts on any charged particle at rest a force F equal to the product of E and the electric charge Q of the particle:

$$F = QE$$

[SOURCE: IEC 60050-121:1998 [10], 121-11-18]

3.10**internal electric field**

E_{ind}

<in a body> internal electric field induced inside the body as a result of exposure to electric, magnetic, or electromagnetic fields

3.11**peak internal electric field**

pE_{ind}

peak instantaneous internal electric field induced inside the body as a result of exposure to electric, magnetic, or electromagnetic fields that is the maximum local, maximum cube averaged (cube length 2 mm) E-field or maximum line averaged (5 mm) E-field

3.12**peak instantaneous internal electric field**

$pE_{\text{ind,inst}}$

maximum internal electric field in time-domain for modulated signals calculated by the multiplication of pE_{ind} with the square root of the peak-to-average ratio

3.13**magnetic field strength**

vector quantity obtained at a given point by subtracting the magnetization M from the magnetic flux density B divided by the magnetic constant μ_0

[SOURCE: IEC 60050-121:1998 [10], 121-11-56, modified – The symbol, formula and Notes to entry have been deleted.]

3.14

magnetic flux density

vector field quantity $\mathbf{B}$ which exerts on any charged particle having velocity $\mathbf{v}$ a force $\mathbf{F}$ equal to the product of the vector product $\mathbf{v} \times \mathbf{B}$ and the electric charge Q of the particle:

$$\mathbf{F} = Q \mathbf{v} \times \mathbf{B}$$

Note 1 to entry: The divergence of $\mathbf{B}$ is zero at all points:

$$\text{div } \mathbf{B} = 0$$

Note 2 to entry: The magnetic flux density $\mathbf{B}$ is sometimes called magnetic field, risking confusion with the magnetic field strength $\mathbf{H}$.

[SOURCE: IEC 60050-121:1998 [10], 121-11-19]

3.15

contact current

<for human body> current flowing between the body and an object at a different electric potential (i.e. when either the body or the object is charged by an EMF)

Note 1 to entry: This is the localized current flow into the body, usually the hand, for a light brushing contact.

[SOURCE: ICNIRP 1998 [4]]

3.16

current density

amount of current per unit cross-section area

Note 1 to entry: Current density is related to the internal electric field by Ohm's Law:

$$\mathbf{J} = \sigma \mathbf{E}$$

where σ is the electrical conductivity of the medium.

[SOURCE: ICNIRP 1998 [4]]

3.17

specific absorption rate

SAR

measure of the rate at which energy is absorbed by the human body when exposed to a radio frequency electromagnetic field

Note 1 to entry: The SAR in the tissue or tissue-equivalent medium can be determined by the rate of temperature increase or by E-field measurements, according to the following formulas:

$$SAR = c_h \left. \frac{\partial T}{\partial t} \right|_{t=0}$$

$$SAR = \frac{\sigma E^2}{\rho}$$

where

SAR is the specific absorption rate in watts per kilogram (W/kg);

E is the RMS value of the electric field strength in the tissue medium in volts per metre (V/m);

σ is the electrical conductivity of the tissue medium in siemens per metre (S/m);

ρ is the mass density of the tissue medium in kilograms per cubic metre (kg/m³);

c_h is the specific heat capacity of the tissue medium in joules per kilogram per kelvin, J/(kg K);

$\left. \frac{\partial T}{\partial t} \right|_{t=0}$ is the initial time derivative of temperature in the tissue medium in kelvin per second (K/s).

[SOURCE: IEC/IEEE 62209-1528:2020, 3.51]

3.18

peak spatial-average SAR

psSAR

maximum average SAR within a local region based on a specific averaging volume or mass

Note 1 to entry: Typical averaging masses are 1 g or 10 g of tissue.

Note 2 to entry: The shape of the averaging volume is cubic as specified in [3] and [8].

3.19

magnitude

<of a vector> for any vector U , non-negative scalar, usually denoted by $|U|$, equal to the non-negative square root of the scalar product or, in the case of a complex vector, of the Hermitian product of the vector by itself

[SOURCE: IEC 60050-102:2008 [11], 102-03-23, modified – The Notes to entry have been deleted.]

3.20

gradient

vector ∇f associated at each point of a given space region with a scalar f , having a direction normal to the surface on which the scalar field has a constant value, in the sense of increasing value of f , and a magnitude equal to the absolute value of the derivative of f with respect to distance in this normal direction

Note 1 to entry: The scalar f can refer to a component of a vector field.

[SOURCE: IEC 60050-102:2008 [11], 102-05-19, modified – Note 1 to entry has been replaced; Notes 2 and 3 to entry have been deleted.]

3.21

phantom

physical model with an equivalent human anatomy and comprised of a tissue-equivalent medium with dielectric properties specified in this document

[SOURCE: IEC/IEEE 62209-1528:2020, 3.39]

3.22

complex relative permittivity

<under sinusoidal conditions in a medium> complex quantity $\underline{\varepsilon}_r$ that linearly relates the electric flux density and the electric field strength

$$\underline{D} = \varepsilon_0 \underline{\varepsilon}_r \underline{E}$$

where

$\underline{D}$ is the electric flux density phasor,

$\underline{E}$ is the electric field strength phasor;

ε_0 is the dielectric permittivity of free space

Note 1 to entry: The complex relative permittivity is generally frequency dependent. For an isotropic medium the complex relative permittivity is a scalar; for an anisotropic medium it is a tensor.

Note 2 to entry: Generally $\underline{\varepsilon}_r$ is expressed as $\underline{\varepsilon}_r = \varepsilon_r' - j\varepsilon_r''$ where ε_r' is the real relative permittivity and ε_r'' is the dielectric loss index. In the context of this document, it represents the conductivity of free charges in the exposed tissue or tissue-equivalent medium $\sigma = \omega \varepsilon_0 \varepsilon_r''$.

[SOURCE: IEC 60050-121:2019 [10], 121-12-14, modified – reworded definition for replacing the term in context.]

3.23

device under test

DUT

device that is tested in accordance with the procedures specified in this document

3.24

post-processing

computation of physical quantities based on the reading of the results of a measurement or on the output of computational software

Note 1 to entry: Examples of the computation are the conversion of currents or voltages obtained from a calibrated probe into electric field strength or magnetic field strength or the averaging of field quantities, e.g. the local SAR distribution for assessment of psSAR.

3.25

wireless power transfer

WPT

transmission of electrical energy by electric, magnetic, or electromagnetic fields without wire or cable connection between the energy source and the load for wireless charging

4 Symbols and abbreviated terms

4.1 Physical quantities

The internationally accepted SI units are used throughout this document.

Symbol	Quantity	Unit	Dimensions
B	Magnetic flux density	tesla	T
c_h	Specific heat capacity	joule per kilogram per kelvin	J/(kg K)
E	Electric field strength	volt per metre	V/m
f	Frequency	hertz	Hz
H	Magnetic field strength	ampere per metre	A/m
J	Current density	ampere per square metre	A/m ²
P	Average (temporal) absorbed power	watt	W
T	Temperature	kelvin	K
ε	Permittivity	farad per metre	F/m
λ	Wavelength	metre	m
σ	Electric conductivity	siemens per metre	S/m

NOTE In this document temperature is quantified in degrees Celsius, as defined by: $T(^{\circ}\text{C}) = T(\text{K}) - 273,15 \text{ K}$.

4.2 Constants

Symbol	Physical constant	Magnitude
η_0	Intrinsic impedance of free space	$120\pi \Omega$ or 377Ω
ε_0	Permittivity of free space	$8,854 \times 10^{-12} \text{ F/m}$
μ_0	Permeability of free space	$4\pi \times 10^{-7} \text{ H/m}$
j	Imaginary unit	1

4.3 Abbreviated terms

BEM	boundary element method
B-field	magnetic flux density
BR	basic restriction
DRL	dosimetric reference limit
DUT	device under test
E-field	electric field strength
ERL	exposure reference level
EV	electric vehicle
FDFD	finite difference frequency domain
FDTD	finite difference time domain
FEM	finite element method
FIT	finite integration technique
GGSM	generic gradient source model
H-field	magnetic field strength
IM	impedance method
MFS	magnetic field source

MQS	magneto-quasi-static
MoM	method of moments
psSAR	peak spatial-average SAR
pE_{ind}	peak instantaneous internal electrical field
RF	radio frequency
RL	reference level
RMS	root mean square
RPA	remotely piloted aircraft
RSS	root sum square
SAR	specific absorption rate
SPFD	scalar potential finite difference
WPT	wireless power transfer (system)

5 Assessment procedures

5.1 General

Compliance assessment is to be carried out in terms of specific absorption rate (SAR), internal electric fields, or current density; and contact currents. Figure 1 presents a flowchart for the assessment procedure for evaluating both the direct effects (internal electric field, current density, or SAR) and contact currents of electromagnetic fields generated by WPT systems. The assessment methods for these quantities are different.

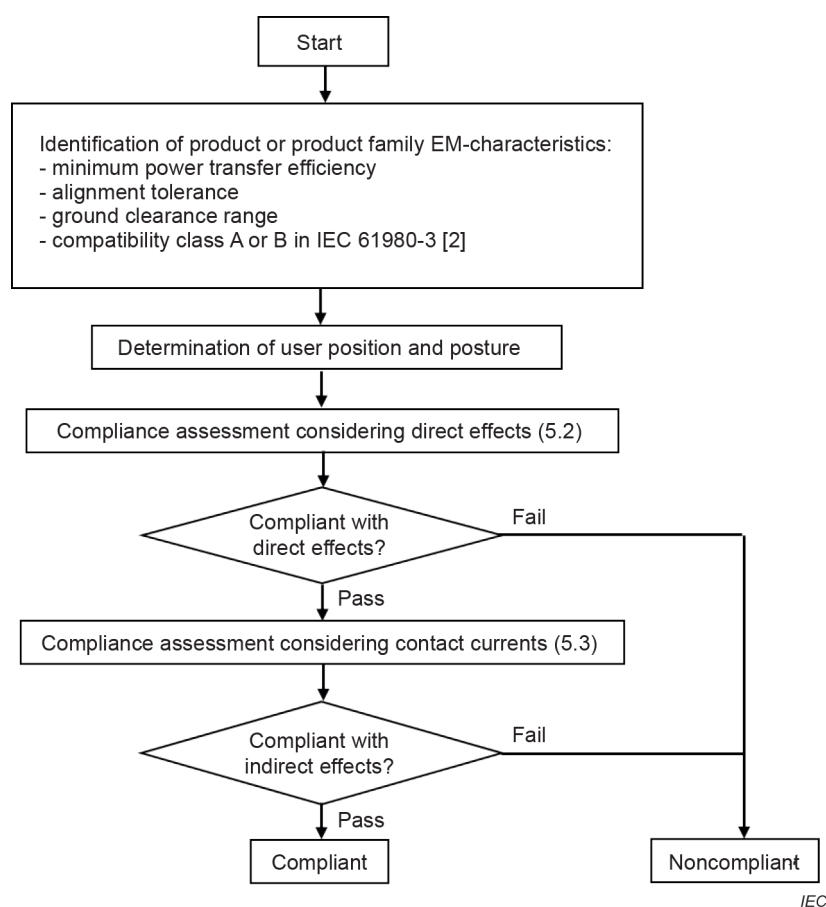


Figure 1 – Flowchart for the assessment procedure

5.2 Compliance assessment considering direct effects

5.2.1 General

Subclause 5.2 provides basic assessment methods considering the direct effects of electromagnetic fields [3], [4], [5], [8]. Evaluations are made in terms of either the incident fields or of the induced fields according to the averaging methods as specified in [3], [4], [5], [8]. This document specifies the implementation of appropriate averaging techniques (Annex G), and references to documents where these are specified. The following list and Figure 2 specify four tiers to assess exposure from WPT and demonstrate conformity with the basic restrictions or reference levels. Any of the four tiers may be selected. They require increasing technical effort and yield decreasing levels of conservativeness.

- Tier 1: The evaluation of the exposure based on coil current is straightforward but is the most conservative one (5.2.2). The Tier 1 method does not evaluate the exposure from the incident electric field.
- Tier 2: Assessment of the incident fields and comparison against reference levels is specified in 5.2.3.
- Tier 3: The incident magnetic fields can be applied to conservatively evaluate the induced fields for comparison with reference levels or basic restrictions using the gradient-source-based method specified in 5.2.4.
- Tier 4: Direct evaluation of the fields induced in the body or in a body phantom using measurement or computational methods is specified in 5.2.5.

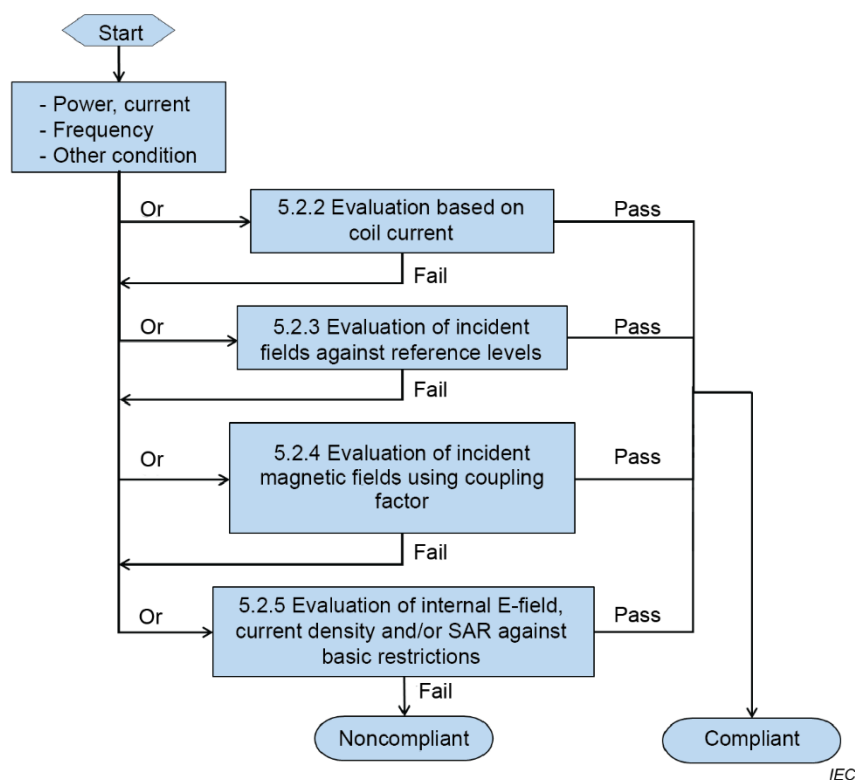


Figure 2 – Flowchart for the assessment procedure considering direct effects

5.2.2 Tier 1: Evaluation based on coil current

According to Clause A.1, the maximum permissible coil current I_{mp} can be approximated by evaluating the magnetic flux density of a circular coil at a fixed distance of 10 mm and comparing it against the reference level B_{RL} . For a circular coil with radius r , I_{mp} can be estimated using Formula (1):

$$I_{mp}(r) = \frac{2\pi d_0 B_{RL}}{\mu_0 n} e^{-\frac{r_0}{r}} \quad (1)$$

where

- r is the maximum coil radius (or half of the largest dimensions of non-circular coils);
- d_0 is the fixed distance of 10 mm at which the magnetic flux density is compared to B_{RL} ;
- B_{RL} is the reference level of magnetic flux density;
- μ_0 is the free space permeability, $4\pi \times 10^{-7}$ H/m;
- n is the number of turns of the coil;
- r_0 is a constant of 21,5 mm.

NOTE The value of r_0 was determined from a curve fit to the numerical solution of the Bio-Savart equation.

The measured coil current shall be less than the calculated I_{mp} .

The approximation of Formula (1) is conservative for fields in the plane of the coil outside its circumference if the diameter of the coil windings is smaller than d_0 . It cannot be applied within the circumference of the coil. Above and below the plane of the coil, the fields can be assumed to decay monotonically.

If it cannot be demonstrated that approximation of Formula (1) is conservative for the DUT or the evaluation distance is different from d_0 , then results in accordance with 5.2.2 are not valid even if the coil current is lower than the I_{mp} .

Because it can be shown that the B-field decays monotonically as a function of the distance from the coil for any coil radius r , results from Formula (1) can be regarded as conservative for distances from the coil larger than d_0 within the uncertainty of the applied approximation (Clause A.1).

Annex A provides additional information regarding exposure evaluations using approximations.

5.2.3 Tier 2: Evaluation of incident fields against reference levels

Both incident electric and magnetic field strengths are assessed at different distances including the closest distance, and in all directions where the DUT can operate near people. The assessment can be carried out using measurement or computational methods or basic calculations.

Depending on the applicable exposure guidelines:

- Above 100 kHz, the incident fields shall be spatially averaged.
- Below 100 kHz, incident fields can be averaged in space over the body or as specified in Annex I.
- Different reference averaging-time periods can be specified for proper assessment of the incident fields against reference levels.

NOTE 1 Additional details on when spatial averaging is permitted can be specified in the exposure guidelines or in national regulations.

NOTE 2 For example, Annex I and Annex J introduce an averaging method for WPT EV systems for persons standing next to the EV on a non-conductive ground. The method is based on the three-point method introduced in [12], which shows that the approach is conservative for measurement distances larger than 20 cm and frequencies up to 100 kHz.

NOTE 3 In some exposure guidelines, the electric and magnetic field strengths are assessed with both an instantaneous (to prevent nerve stimulation) and six minutes reference averaging-time periods.

5.2.4 Tier 3: Evaluation of incident magnetic fields using coupling factor

5.2.4.1 General

The reference levels of electromagnetic fields are derived from the maximum coupling conditions between the fields and the human body, i.e. the maximum induced quantities in the human body exposed to the uniform electromagnetic fields.

Coupling factors for magnetic field exposure can be applied for the incident field to mitigate the over-estimation and still warrant conservativeness with respect to the basic restrictions, based on Formula (2).

$$QoI = k \times F \quad (2)$$

where

QoI is the quantity of interest to be compared in terms of the basic restrictions;

k is a generic coupling factor that shall be derived according to 5.2.4.2 for specific coil configurations and evaluation distances, or according to 5.2.4.3 for the generic gradient source model (GGSM);

F is the measured incident field at the location of the evaluation.

Two methods are specified in this document as options:

- the coupling factor k_L is estimated from localized exposure scenarios as specified in 5.2.4.2;
- the generally applicable coupling factor k_G is characterized with respect to the normalized incident magnetic field gradient at the location of the exposure evaluation by applying the GGSM method [13] (5.2.4.3).

NOTE The method specified in 5.2.4.2 uses coupling factors for the maximum induced current density, electric field, and SAR. The calculation results of coupling factors are listed in IEC TR 62905 [14]. The GGSM method specified in 5.2.4.3 specifies coupling factors for all basic restrictions of [3], [4], [5], and [8].

The methods of 5.2.4.2 or 5.2.4.3 can be applied only under the following conditions:

- a) The induced quantities such as internal electric field, current density, or SAR caused by incident electric fields are negligible compared to those induced by the incident magnetic fields.
- b) The whole-body average SAR is negligible compared to the local exposure limits, and the contribution of the psSAR or the internal electric field is dominant.
- c) Evaluation conditions such as frequency, field distribution, distance to human body, etc., are identical to those used in the derivation of the coupling factor. Otherwise the uncertainty due to the deviation shall be determined following the methodologies of [12], [15] for 5.2.4.2, and of [13] for 5.2.4.3.
- d) Quasi-static conditions (7.2) apply for the incident magnetic field.

In addition to the variables specified for Formula (2), descriptions of the symbols used in the formulas of 5.2.4.2 and 5.2.4.3 are consolidated in Table 1.

Table 1 – List of symbols used in the formulas of 5.2.4.2 and 5.2.4.3

Symbol	Description	Constant	Formula (#)
B_0	is a constant normalization magnetic flux density of 6,25 μT [5]	yes	(14)
B_1	is a constant normalization magnetic flux density of 2,04 μT [16]	yes	(15)
$B_{\text{inc, meas}}$	is the local magnetic flux density measured by a field probe or evaluated computationally	no	(12), (13)
B_{max}	is the maximum magnetic flux density in the absence of a person at a point inside the bounding box within which the person could be located	no	(3), (4), (5)
$B_{\text{max}}(G_n)$	is the maximum measured magnetic flux density at the human body surface when the field gradient is G_n	no	(6)
B_{RL}	is the applicable reference level of the magnetic flux density B or a fixed normalization value, e.g. B_0	yes	(3), (4), (5), (6), (12), (13)
c_3	constant = 1 m^{-1}	yes	(7), (8), (9), (10), (11)
E_{BR}	is the basic restriction of internal electric field strength	yes	(4), (12)
E_{max}	is the maximum internal electric field strength induced in the human body	no	(4)
F	is the measured incident field at the location of the evaluation	no	(2)
f_0	is a constant normalization frequency of 100 kHz	yes	(14), (15)
G_n	normalized local B-field gradient perpendicular to the skin surface of the exposed body	no	(6), (7), (8), (9), (10), (11)
$G_{n, \text{meas}}$	measured normalized local B-field gradient	no	(12), (13), (14), (15)
J_{BR}	is the basic restriction of current density	yes	(3), (13)
J_{max}	is the maximum current density induced in the human body	no	(3)
k	is a generic coupling factor that shall be derived according to 5.2.4.2 for specific coil configurations and evaluation distances, or according to 5.2.4.3 for the generic gradient source model (GGSM)	yes	(2)
$k(G_{n, \text{meas}})$	is the value of the coupling factor calculated for the specific gradient $G_{n, \text{meas}}$	yes	(12), (13)
$k_{B, 10\text{g}}$	is the low frequency magnetic flux density coupling coefficient for psSAR10g, 2×10^{-4}	yes	(14)
$k_{B, 1\text{g}}$	is the low frequency magnetic flux density coupling coefficient for psSAR1g, $4,6 \times 10^{-5}$	yes	(15)
$k_{G, E; \text{line}}$	coupling factor for induced E-field averaged over a line of 5 mm	yes	(8)
$k_{G, E; \text{cube}}$	coupling factor for induced E-field averaged over a cube of 8 mm ³	yes	(7)



Symbol	Description	Constant	Formula (#)
$k_{G,e; \text{ circle}}$	coupling factor for current density averaged over an area of 1 cm ²	yes	(9)
$k_{G,psSAR10g}$	coupling factor for peak spatial SAR averaged over a cube of 10 g	yes	(10), (14)
$k_{G,psSAR1g}$	coupling factor for peak spatial SAR averaged over a cube of 1 g	yes	(11), (15)
QoI	is the quantity of interest to be compared to the basic restrictions	no	(2)
QoI_{BR}	is the basic restriction of the induced metric QoI	yes	(6)
$QoI_{\max}(G_n)$	represents the maximum of the induced quantity of interest (electric field strength, current density, or SAR) in the body when the field gradient is G_n	no	(6)
SAR_{BR}	is the basic restriction for psSAR	yes	(5), (14), (15)
$SAR_{\max}$	is the maximum psSAR induced in the human body	no	(5)

5.2.4.2 Coupling factor k_L based on the analysis of localized exposure scenarios

The coupling factor for the current density $k_{L,J}$ is determined using Formula (3).

$$k_{L,J} = \frac{J_{\max}/B_{\max}}{J_{BR}/B_{RL}} \quad (3)$$

where the symbols are specified in Table 1 of 5.2.4.1.

NOTE 1 The coupling factor was originally introduced for exposure assessments in low frequencies in IEC TR 62905:2018 [14], IEC 62226-2-1 [17] and IEC 62311:2007 [18], then amended in IEC 62233 [19].

Using this coupling factor, the measured magnetic field strength, multiplied by the coupling factor, is directly compared to the reference level.

The concept of exposure assessment using coupling factor is expanded to the internal electric field and SAR with Formula (4) and Formula (5) ([12], [15]).

$$k_{L,E} = \frac{E_{\max}/B_{\max}}{E_{BR}/B_{RL}} \quad (4)$$

$$k_{L,SAR} = \frac{\sqrt{SAR_{\max}}/B_{\max}}{\sqrt{SAR_{BR}}/B_{RL}} \quad (5)$$

where the symbols are specified in Table 1 of 5.2.4.1.

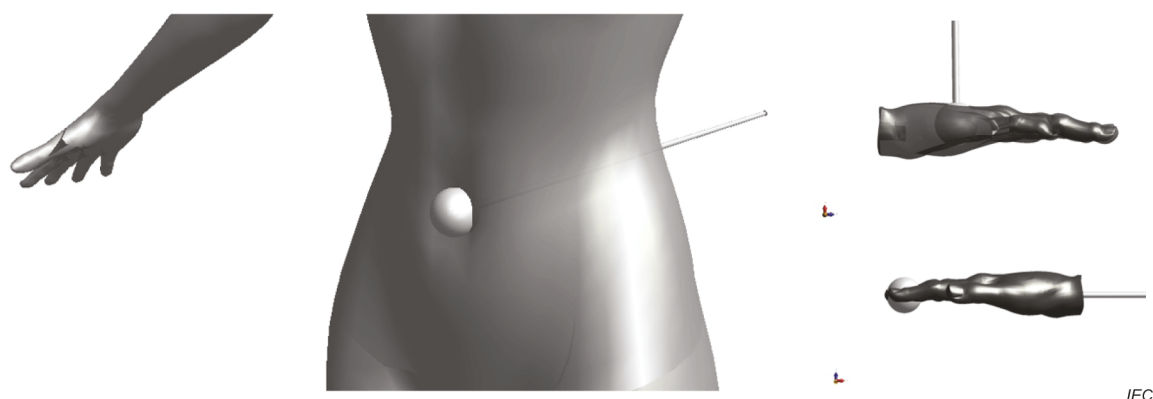
E_{BR} and B_{RL} depend on frequency. The coupling factor k_L is multiplied by the measured magnetic fields to consider the effects of field non-uniformity. Conservative coupling factors k_L are typically determined by simulation with the ratio of the induced quantities to the incident magnetic field ([12], [15]), e.g. for a specific frequency, field distribution, distance to human body and for different anatomies. Based on such an evaluation, conservative values of the coupling factors can be derived, the uncertainty of which shall be estimated and documented following the methodology applied in [12], [15]. Appropriate methods for the computational modelling of the coil and the exposed person are specified in Clause 7.

NOTE 2 See also Annex F and Annex G. Annex H specifies procedures for validation of the computational results.

Exposure in terms of the internal electric field strength, current density, or psSAR is then assessed by comparing the multiplied value to the reference level of the magnetic fields only.

5.2.4.3 Coupling factor k_G based on the GGSM method

A generally applicable coupling factor $k_G(G_n)$ is introduced using the GGSM of [13] to mitigate over-estimation by incident magnetic field assessment in strong field gradients that are common in the vicinity of WPT systems. The method is conservative for real sources because the coupling factors are derived from a 2D gradient model, whereas real sources have gradients in 3D. However, it requires evaluation of both the local field amplitude and the local gradient in the normal direction with respect to the exposed body, as indicated in Figure 3.



NOTE In each view the sphere with attached shaft symbolizes example locations of a field probe to measure the field amplitude for evaluating the gradient at the reference location (body surface).

Figure 3 – The gradient G_n is determined at the surface and normal to the surface, i.e. in the direction of the axis shown

The method applies the following quantities of the incident magnetic field at the location of the exposed body:

- frequency;
- magnetic flux density (B_{xyz}) at the skin surface of the exposed body;
- normalized local B-field gradient G_n (T/m/T) perpendicular to the skin surface of the exposed body.

NOTE 1 Because the coupling factors only depend on the spatial properties of the incident magnetic field and not on its amplitude, they are given for the field gradient normalized to the local field amplitude and expressed in units of T/m/T.

The GGSM coupling factors have been determined for a wide range of 2D gradients (i.e. uniform in one direction) from a generic magnetic field source comprised of two parallel line currents with equal amplitudes but 180° phase difference and different separation distances as specified in [13]. This source has a magnetic flux density B and a normalized B-field gradient G_n that depend on a position relative to the GGSM source and the magnitude of the source current. The fields induced in the exposed body are normalized to the maximum B-field amplitude and normalized B-field gradient at the skin surface. Based on a large-scale computational study including human body models of different ages, heights, and weights [13], worst-case exposure conditions (i.e. the highest induced field quantities) were determined for each normalized B-field gradient and frequency analysed. The coupling factors $k_G(G_n)$ are determined using Formula (6):

$$k_{G,QoI} = \frac{QoI_{\max}(G_n)/B_{\max}(G_n)}{QoI_{BR}/B_{RL}} \quad (6)$$

where the symbols are specified in Table 1 of 5.2.4.1.

NOTE 2 The coupling factors were determined for the heterogeneous models of the Virtual Population (ViP) ([20], [21]) and can be regarded as conservative, i.e. they already include the enhancement factors caused by tissue boundaries.

The coupling factors of each specific induced metric specified in Annex G according to the guidelines, i.e. [3], [4], and [5], are expressed by physics-based approximations as functions of the gradient of the magnetic flux density, i.e. Formula (7), Formula (8), Formula (9), Formula (10), and Formula (11):

$$k_{G,E; \text{cube}}(G_n) = \frac{1}{\left[1 + 6,5 \times 10^{-6} \times (G_n/c_3)^{5,8}\right]^{(1/5,8)}} \quad (7)$$

$$k_{G,E; \text{line}}(G_n) = \frac{1}{\left[1 + 4 \times 10^{-11} \times (G_n/c_3)^{6,6}\right]^{(1/6,6)}} \quad (8)$$

$$k_{G,J; \text{circle}}(G_n) = \frac{1}{\left[1 + 4 \times 10^{-3} \times (G_n/c_3)^{2,9}\right]^{(1/2,9)}} \quad (9)$$

$$k_{G,\text{psSAR } 10\text{g}}(G_n) = \left(\frac{1}{\left[1 + 2 \times 10^{-1} \times (G_n/c_3)^{1,2}\right]^{(1/1,2)}} \right)^2 \quad (10)$$

$$k_{G,\text{psSAR } 1\text{g}}(G_n) = \left(\frac{1}{\left[1 + 2,5 \times 10^{-1} \times (G_n/c_3)^{1,1}\right]^{(1/1,1)}} \right)^2 \quad (11)$$

where the symbols are specified in Table 1 of 5.2.4.1.

The trend of the approximation formulas as a function of the gradient is shown in Figure 4. The method has been analytically verified (see Appendix of [13]). For any other specific induced metrics or basic restrictions, the process specified in [13] shall be repeated.

To assess exposure in terms of the basic restrictions, the magnetic flux density (B_{meas}) and the normalized field gradient ($G_{n,\text{meas}}$) of the DUT are evaluated at any closest accessible location from the source using equipment that provides both values, or that measures the field using a magnetic vector-field meter at separate locations and derives the gradient vectors of the vector field components of the magnetic field. Here the measured magnetic field gradient is determined at the reference location according to Figure 3.

The internal electric field, current density and SAR can then be predicted by Formula (12), Formula (13), Formula (14), and Formula (15) [13]:

$$E = k_{G,E} (G_{n,\text{meas}}) \times E_{\text{BR}} \times \frac{B_{\text{inc,meas}}}{B_{\text{RL}}} \quad (12)$$

$$J = k_{G,J} (G_{n,\text{meas}}) \times J_{\text{BR}} \times \frac{B_{\text{inc,meas}}}{B_{\text{RL}}} \quad (13)$$

$$psSAR_{10g} = k_{G,psSAR10g} (G_{n,\text{meas}}) \times SAR_{\text{BR}} \times \left(\frac{B_{\text{inc,meas}}}{B_0} \right)^2 \times \left(\frac{f}{f_0} \right)^2 \times k_{B,10g} \quad (14)$$

$$psSAR_{1g} = k_{G,psSAR1g} (G_{n,\text{meas}}) \times SAR_{\text{BR}} \times \left(\frac{B_{\text{inc,meas}}}{B_1} \right)^2 \times \left(\frac{f}{f_0} \right)^2 \times k_{B,1g} \quad (15)$$

where the symbols are specified in Table 1 of 5.2.4.1.

The formulas presented in this 5.2.4.3, together with Figure 4, quantify the reduction of the induced field quantities in the presence of a magnetic gradient field with a coupling factor k of less than 1.

NOTE 3 An example is a case of two magnetic field sources that operate at an identical frequency and whose magnetic fields at the measurement point are equal. One of them has a uniform magnetic field distribution ($G_n = 0$), and the magnetic field gradient of the other one is different from zero (the magnetic field decreases in the direction of the human body). In both cases, E_{BR} , B_{RL} , and $B_{\text{inc,meas}}$ are the same, but $k_{G,E} = 1$ for the uniform field and $k_{G,E} < 1$ for the field with gradient. This makes the induced electric field E of the non-zero-gradient smaller than that of the no-gradient case.

NOTE 4 For large field gradients, results of Formula (7) through Formula (11) approach zero. This can be explained as that the fields inside the volume of the body being exposed approach zero as the gradient approaches infinity.

The method may be applied using the following steps:

- determine the incident B-field at all locations at which a body can be exposed to the field of the source and determine the lowest gradient that is normal to the surface of the body (see Figure 3);
- compute the maximum induced fields according to Formula (12), Formula (13), Formula (14), and Formula (15) for all locations.

If the maximum induced field is lower than the applicable basic restriction, the source is compliant.

For the application of this method the conditions listed in 5.2.4.1 shall apply. An additional condition is that the incident magnetic field is dominated by a single source. For conditions different from these, the uncertainty should be determined following the methodology of [13].

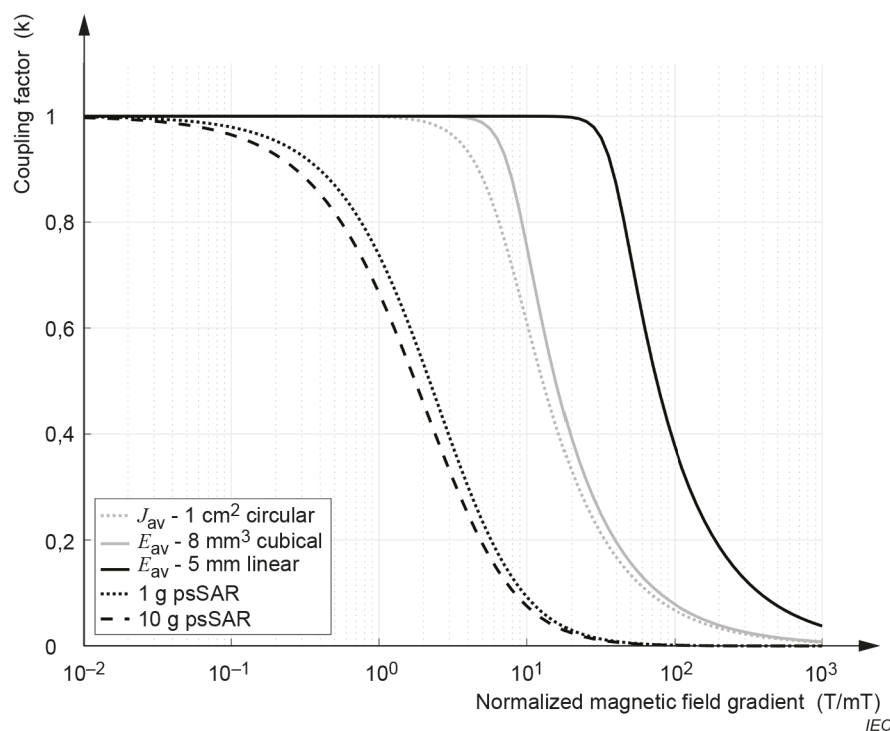


Figure 4 – Coupling factors k of Formula (7) through Formula (11) as a function of the normalized magnetic field gradient [13]

5.2.5 Tier 4: Evaluation of internal E-field, current density, or SAR against basic restrictions

SAR can be measured using a small probe in a liquid-filled phantom which is exposed to the electromagnetic fields of the DUT. The measured values can be directly compared with basic restrictions. The evaluation and validation procedures are specified in IEC/IEEE 62209-1528. The assessment method is specified in detail in 6.2.

NOTE The frequency range of IEC/IEEE 62209-1528:2020 is from 4 MHz to 10 GHz.

Internal electric field, current density, or SAR may be evaluated analytically or by simulation using anatomical models of the human body or models of laboratory phantoms for comparison with basic restrictions. The simulation model and methodology shall be validated as part of the evaluation process. The validation procedure and results shall be documented. The assessment method based on simulation is specified in detail in Clause 7.

5.3 Exposure assessment of contact currents

Contact currents are the only indirect effects considered in this document. They flow into a biological body via a physical contact with the electrode or other source of current.

Contact currents can be assessed either by direct measurement or by measurement of the incident electromagnetic fields.

The following considerations shall be applied to assess exposure by incident electromagnetic fields.

- The objects exposed to H-fields are grounded (Figure 5).
- The contact currents as a function of the incident magnetic field strength are calculated with Formula (16), which is derived from Faraday's law.

$$I_c = H_{inc} \frac{2\pi f \mu_0 S}{Z} \quad (16)$$

where

- I_c is contact current (RMS);
- H_{inc} is the RMS of the incident H-field;
- f is the frequency of the incident H-field;
- μ_0 is the permeability of free space;
- Z is the impedance of the human body (see 6.3);
- S is the area of the loop such as 1,5 m × 0,5 m.

Contact currents may be calculated with Formula (16) or measured directly as specified in 6.3. Contact currents can be measured directly using a contact current meter with an ungrounded or grounded metal object placed in the vicinity of the DUT. Figure 5 shows exposure situations for ungrounded and grounded metal objects.

Figure 6 presents a flowchart for the assessment procedure. Details of measurement conditions, measurement setup, etc., are summarized in IEC TR 63167 [22]. The assessment method is specified in detail in 6.3.

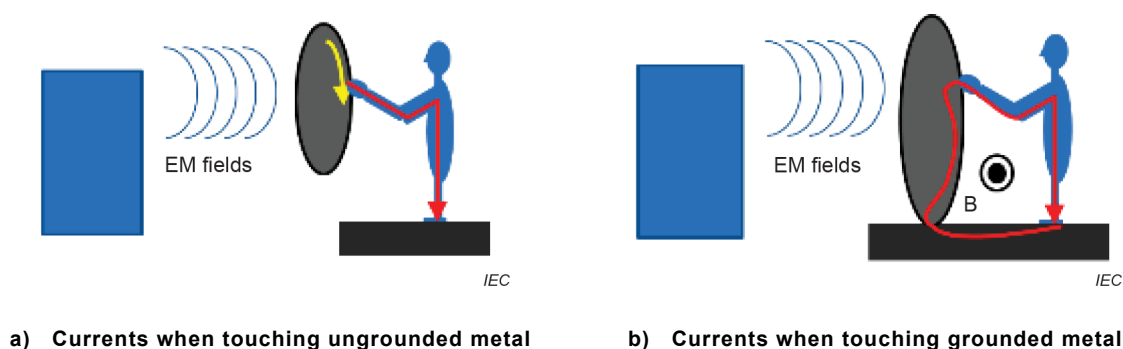


Figure 5 – Two exposure situations for ungrounded and grounded metal objects

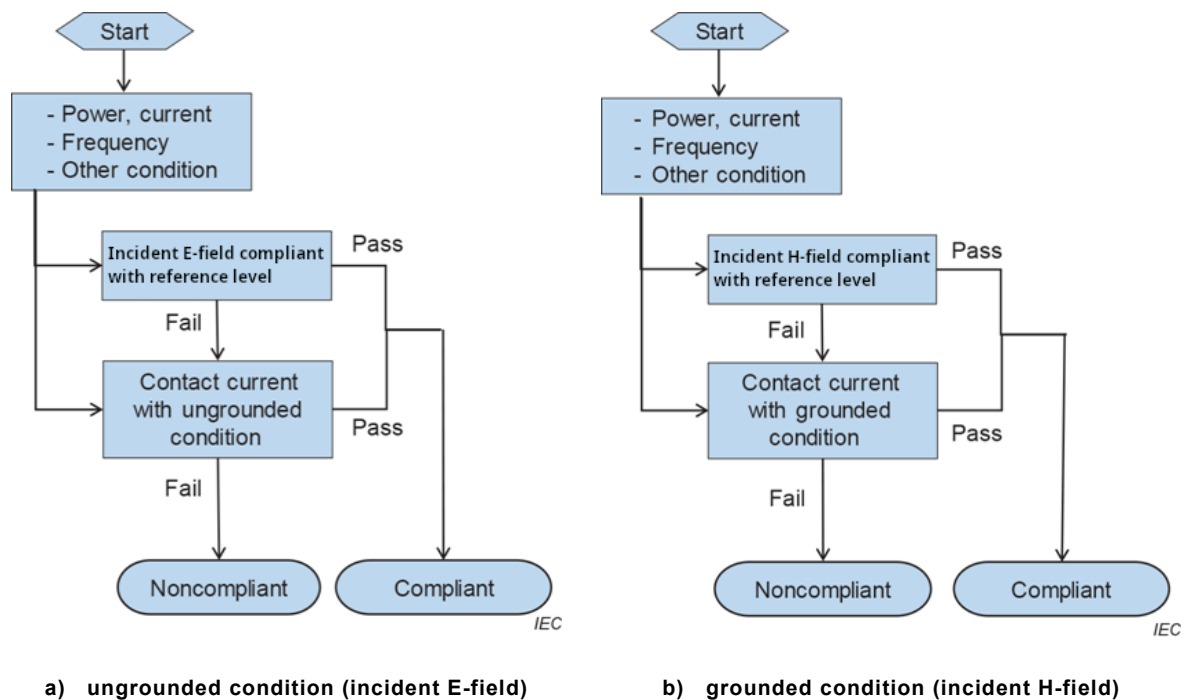


Figure 6 – Flowchart for assessment procedures for contact currents

6 Measurement methods

6.1 Incident fields

6.1.1 General

IEC 61786-1:2013 and IEC 61786-1:2013/AMD1:2024 and IEC 61786-2:2014 specify electromagnetic field basics, measurement methods, and requirements of the measurement equipment at frequencies below 100 kHz. In addition, IEC 62233 [19] and IEC 62311 [18], [23] specify electromagnetic field measurements for the purpose of exposure assessment. This document specifies the requirements for measurement equipment for electromagnetic fields from 3 kHz to 30 MHz.

- Equipment for measuring electric fields, magnetic fields, and magnetic field gradient is specified in 6.1.2.1, 6.1.2.2, and 6.1.2.3, respectively. The calibration methods are specified in Annex B.
- The validation method is specified in Annex C.
- Both incident electric and magnetic field vectors are measured at a specific distance, including the closest distance where the DUT may operate near people. These International Standards specify measurement positions based on the size and regular usage of equipment. IEC 62311 [18], [23] specifies it as a "user's usual position." IEC 62233 [19] specifies the measurement position in detail for household appliances and similar apparatus, including inductive battery chargers.
- The measured value may be based on either a time-averaged or the maximum instantaneous value of the target electromagnetic field, depending on the characteristics of the source and the exposure metric being assessed.
- The incident fields are measured in free space without the presence of an exposed body or phantom. The closest measurement distance should be chosen within the range that is accessible to the user.

- f) When the measurement probe comes close to the source and metallic or highly insulated objects, electromagnetic coupling may occur and the probe sensor impedance changes. Due to the finite size of the probe, all field measurements are subject to some amount of spatial averaging. For a loop, the measurement location may be specified as the geometric centre of the loop, but the result will be a function of the average H-field passing through the loop area. The probe sensor should be sufficiently small for improved accuracy of the measurement of the spatial peak of a given field component. Larger probes are prone to larger uncertainties in strong gradient fields. The gradient and isotropy uncertainty shall be properly evaluated with its instrumentation by simulations and by rotating the probe in the gradient field (9.2.7).

NOTE 1 For example, if an isotropic sensor with a loop area of 1 cm² is rotated in a magnetic field with a normalized gradient of 50 T/m/T, which is constant over the area of the loop, the resulting error is less than 1,3 % (rectangular distribution). For a sensor with a loop area of 3 cm², it is less than 5 %.

NOTE 2 Additional guidance on incident field measurement configurations, probe size considerations, etc. is available for example in [24].

- g) In the case of non-uniform incident fields, the measured field values may be spatially averaged over the exposed regions of the body, with the important provision that the basic restrictions on psSAR and induced electric fields are not exceeded. If averaging is applied in the case where one or more of the measured values are above the reference levels, evidence shall be provided that the assessment based on the averaged values does not lead to exceeding the underlying basic restrictions.

NOTE 3 National regulations can differ from the preceding conditions and pose different requirements, e.g. on frequency range, minimum distance, or acceptability of spatial averaging as specified in step g). Incident electric and magnetic field strengths are measured or calculated in accordance with the applicable regulations.

Assessment procedures for WPT, heavy duty EVs, and remotely piloted aircraft are specified in Clause I.1, Clause I.2, and Clause I.3, respectively. Examples of assessment results are provided in Annex J.

If the incident field consists of multiple frequency components (i.e. above the noise floor of the measurement set-up), these contributions shall be considered in the exposure assessment. Summation regimes (schemes) in Annex A of IEC 62311:2019 [23] may be applied to consider spectral contributions.

6.1.2 Equipment

6.1.2.1 Equipment for electric field strength measurements

Electric field strength can be measured with a calibrated antenna or probe that is sensitive to electric fields exclusively (e.g. small dipoles) and connected to the calibrated measurement equipment.

Two types of antennas or probes are used for electric field measurements: single-axis-sensor types, and three-axis-sensor types in which three sensors are arranged perpendicular to each other. IEC 61786-2 recommends using the three-axis type. Using single-axis instruments is possible in some cases, e.g. when the direction of the field is already known (IEC 61786-1) or if the probe is rotated and all three vector components of the field are evaluated. The electric field strength of the three-axis component readings given in Formula (17) applies:

$$E_R = \sqrt{E_x^2 + E_y^2 + E_z^2} \quad (17)$$

where E_x , E_y , and E_z are effective electric field strengths for the three orthogonal axes.

Three-axis electric field measurement equipment has the advantage of measuring the net field E_R regardless of the probe orientation.

In general, measurement equipment for electric field strength is either broadband or narrowband. For both measurement systems, the equipment consists of an antenna or probe (including power source), measurement part, and data recording and processing unit. The following describes the system characteristics and selection method.

- a) A broadband system uses dipole antennas. It is small enough to improve the uniformity of the sensitivity at each frequency. To improve isotropy, three antennas are generally placed orthogonal to one another and each signal or the combined signal is extracted. This type of antenna is suitable for near-field electromagnetic fields of complicated distributions. A broadband isotropic electric field strength meter designed for radio wave protection is typically used for measurement provided that the electric field strength is uniform over the volume of the probe. Field uniformity can be established by checking that the result remains within the expected measurement uncertainty when the probe is rotated to arbitrary orientations. If the electric field is not sufficiently uniform over the volume of the probe at the measurement location, the uncertainty of the measured field at the centre of the probe caused by the electric field gradient shall be assessed. Other systems may include waveform observation equipment, such as diode detectors and oscilloscopes.
- b) A narrowband system is tuned to each target frequency. The measurement system generally has broad measurement frequencies, high receiving sensitivity, and short responses. It is thus also suitable for pulsed waves or multiple sources. However, for the pulsed wave measurement, it is acceptable to measure in the frequency domain where the target frequency range is broader than the passband width of the measurement system. A biconical antenna with broadband characteristics is often used for far-field region because the size of the antenna is not electrically small. A three-axis orthogonal small dipole is suitable for near-field measurements.
- c) For an electromagnetic environmental assessment, a broadband isotropic electric field strength meter is preferred unless the DUT cannot be measured properly with such a device, in which case it shall be replaced by another measurement system.
- d) The incident E-fields are only measured with low uncertainty if the E-field is uniform or has a constant gradient over the entire E-field sensor. Only these types of field sources are relevant for human exposure. Localized E-field sources caused by accumulation of charges have a strong non-uniform gradient up to $1/d^4$, with d as the distance between the field source and the exposed body. However, the fields couple poorly to the body, i.e. < -60 dB [25]. If it can be demonstrated that induced E-field by these local sources that are characteristic of the DUT is much less than 1 % of the compliance limit, it is not necessary to assess the local E-field.

When selecting a measurement system, the required performance, frequency response, and electromagnetic field strength range shall be considered.

6.1.2.2 Equipment for magnetic field strength measurements

Search-coil sensors are often used for a low frequency range of several hundred kilohertz and below. Different frequency bands can be covered by different sensors. In general, the search-coil probe is used for measuring environmental magnetic fields. The coil probe measures the induced voltage associated with the time-varying magnetic flux normal to the plane of the coil, averaged over the coil area. In a field that is uniform over the area of the coil, the induced voltage at the coil terminals is calculated according to Formula (18):

$$V = -j\omega NBS \quad (18)$$

where

V is the induced voltage;

ω is the angular frequency ($= 2\pi f$);

N is the number of turns;

B is the magnetic flux density (normal to the plane of the coil);

S is the cross-sectional area of the coil;

and all quantities are phasors.

The uncertainty of the measured field at the centre of the sensor caused by the gradient or other non-uniformities of the field shall be assessed (see also step f) of 6.1.1).

To achieve better sensitivity, the area and winding number are increased, and the magnetic flux density is enhanced by using ferromagnetic material in the core of the coil. The signal is processed to obtain flat frequency characteristics by an integrator in subsequent stages because the output is proportional to the frequency.

Magnetic field measurement equipment is divided into two classes: single-axis-probes with a single sensor, and three-axis-probes with three orthogonal sensors. IEC 61786-2 recommends using the three-axis type. The magnetic flux density of the three-axis component reading given in Formula (19) applies:

$$B_R = \sqrt{B_x^2 + B_y^2 + B_z^2} \quad (19)$$

where B_x , B_y , and B_z are effective magnetic flux densities for the three orthogonal axes.

Three-axis magnetic field measurement equipment has the advantage of measuring the net field B_R regardless of the probe orientation.

6.1.2.3 Measurement equipment for the magnetic field gradient

The evaluation method based on the coupling factor as a function of the magnetic field gradient, i.e. 5.2.4.3, requires knowledge about the local B-field amplitude and gradient.

The magnetic field amplitude and gradient can be measured by sampling the field spatially in 3D around the location of exposure. The measurement can be performed by a single field sensor or multiple isotropic magnetic field sensors. A probe with a single sensor can be moved, e.g. by a robot system, to measure the field at separate locations of known spatial placement, if the test signal is repetitive and the phase relation can be established. A probe with multiple sensors can be used to measure the gradient directly, i.e. an axial or planar gradiometer [24], [26], [27], [28].

6.2 SAR and pE_{ind}

SAR and pE_{ind} can be measured and directly compared with the basic restrictions of the applicable requirement. The evaluation and validation procedures of internal electric field distribution and SAR measurements for wireless communication devices are specified in IEC/IEEE 62209-1528 (for the frequency range 4 MHz to 10 GHz). Similar measurement procedures can be used for WPT systems within the frequency range of 4 MHz to 30 MHz. In this document, only the deviations required to assess the spatial peak SAR values of WPT systems are specified. Clause B.4 specifies the procedures for probe calibration. Using an isotropic electric field probe, the SAR inside an irradiated body model can be determined. A SAR measurement system consists of a suitable phantom, an electric field probe, and a probe scanning system.

NOTE 1 The frequency limitation is given by the probe technology defined in IEC/IEEE 62209-1528:2020. As soon as other probes become available, the concept can directly be extended to frequencies below 4 MHz.

The physical characteristics of the phantom model (size, shape, electrical properties, etc.) shall simulate the human exposure condition by the DUT. Phantom models such as those specified in IEC/IEEE 62209-1528 may be used. The phantom model shall consist of materials with dielectric properties similar to humans. At a frequency lower than 30 MHz, the electrical properties of the material are dominantly affected by the conductivity of the material. The liquid specified in Table 2 or NaCl solution with the conductivity specified in Table 3 shall be used at frequencies between 0,1 MHz and 30 MHz. The relationship between conductivity and NaCl concentration is described in [29].

NOTE 2 The same requirements as in Table 2 are specified in IEC/IEEE 62209-1528:2020 for 4 MHz to 30 MHz.

The dependency of SAR on phantom properties and size are shown in Annex D. The same media may be used for frequencies below 4 MHz provided that the uncertainty with respect to conservative estimation is determined, considered, and documented (see 9.3.3).

NOTE 3 Reference [29] shows that the dielectric properties of saline solution differ by less than 5 % from their value at 1 GHz when the frequency tends to zero. This is why constant parameters are specified in Table 2 and Table 3 for the entire frequency range from 0,1 MHz to 30 MHz

Table 2 – Dielectric properties of the tissue-equivalent medium liquid

Frequency (MHz)	Real part of the complex relative permittivity, ε'_r (dimensionless)	Conductivity, σ (S/m)
0,1 to 30	55	0,75

**Table 3 – Dielectric properties of the tissue-equivalent medium
NaCl solution of 0,074 mol/L**

Frequency (MHz)	Real part of the complex relative permittivity, ε'_r (dimensionless)	Conductivity, σ (S/m)
0,1 to 30	77	0,75

The DUT shall be operated at the closest accessible distance to the phantom. Due to the high SAR gradients of the induced field, measurements shall be performed with a known uncertainty of the field at the interface of the phantom shell to the liquid. An extrapolation method for the induced field may then be used to determine the SAR peak at the phantom surface. In that case, uncertainty due to the extrapolation method with respect to SAR values shall be included in the uncertainty budget.

NOTE 4 Examples for extrapolation methods are given in Annex E.

Extensions of the area scan or volume scan procedures may be used to determine the location of maximum SAR. Subclause 7.4 of IEC/IEEE 62209-1528:2020 specifies the scanning procedures. The scanning region shall be 20 % larger than the source or as large as the phantom, whichever is smaller. Additional zoom scans shall be measured for secondary peaks if the primary peak is within 2 dB of the SAR limit.

NOTE 5 Refer to applicable national regulations for the specified SAR limit.

The zoom scan parameters shall be those of Table 4 of IEC/IEEE 62209-1528:2020 and shall meet the requirements given in steps 3 through 5 of 7.4.2 of IEC/IEEE 62209-1528:2020.

The pE_{ind} can be derived from the measured local maximum SAR value in the phantoms standardized in IEC/IEEE 62209-1528 using Formula (20):

$$pE_{\text{ind}} = \sqrt{pSAR \times \frac{\rho}{\sigma}} \quad (20)$$

where

pE_{ind} is the maximum E-field at the phantom surface;

$pSAR$ is the local specific absorption rate;

σ is the conductivity;

ρ is the equivalent tissue density of 1 000 kg/m³ to account for the actual density of brain or body tissue.

The maximum instantaneous E-field value at the phantom surface $pE_{\text{ind,inst}}$ is then given by Formula (21):

$$pE_{\text{ind,inst}} = pE_{\text{ind}} \sqrt{PAR} \quad (21)$$

where

$pE_{\text{ind,inst}}$ is the maximum instantaneous E-field at the phantom surface;

pE_{ind} is given in Formula (20);

PAR is the signal peak power to average ratio.

6.3 Contact currents

6.3.1 General

Subclauses 6.3.2 and 6.3.3 describe equipment and measurement protocol for the assessment of contact currents. IEC TR 63167 [22] provides general information related to human exposure to electric, magnetic, and electromagnetic fields including assumed situations and a proposed method for measuring contact currents.

6.3.2 Equipment

An electrical equivalent circuit human body impedance model, which is frequency dependent, is used to measure contact currents. Figure 7 shows an equivalent circuit, as depicted in IEC 60990 [30], which simulates a human body. Several other studies have evaluated human body impedance and proposed more precise equivalent circuits using arrangements of resistors and capacitors [31], [32], [33], [34]. The frequency characteristics of human body impedance with grasping conditions are compared in Figure 8. The impedance depends on area of contact and usually the impedance for grasping contact is lower than that for point contact.

NOTE The assumed conditions of contact are "point contact" (area not specified) in ICNIRP guidelines [4], [5], and "touch contact" with a contact area of 1 cm² and "grasping contact" with a contact area of 15 cm² for IEEE safety guidelines [8]).

Figure 9 shows the configuration of a contact current meter. The contact terminal is made of metal. If the target point of the contact part is painted, a 10 cm × 20 cm metal foil should be attached to it to simulate a hand palm contact, according to IEC 60990 [30]. A metal board equivalent to the size of a foot is used as the ground electrode. The contact current is measured from the current flowing through the equivalent circuit of the human inside the meter.

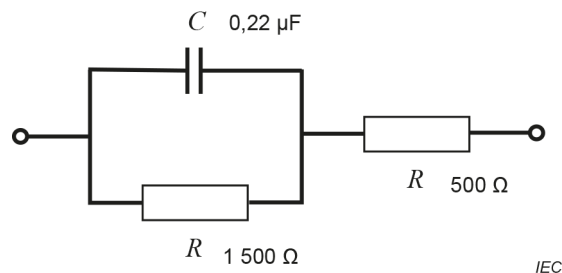


Figure 7 – Human body equivalent circuit proposed in IEC 60990 [30]

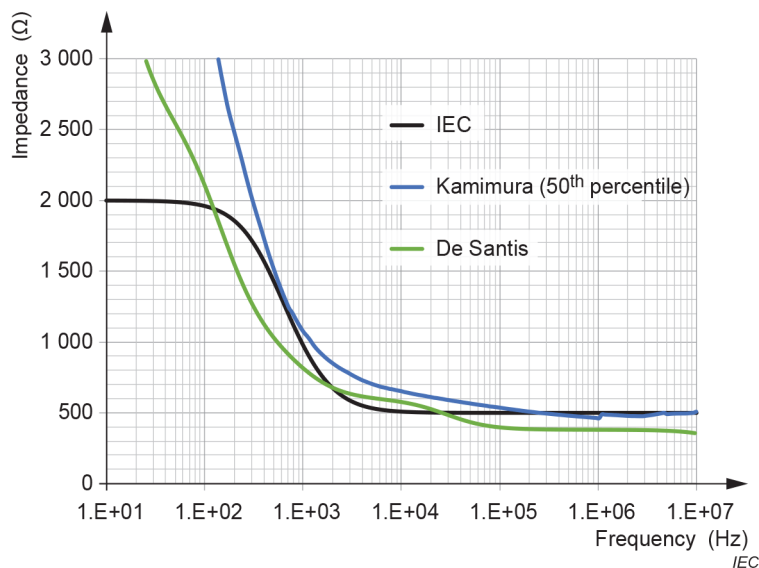


Figure 8 – Impedance frequency characteristics of adult male and equivalent circuits proposed in IEC 60990 [30] and evaluated values [31], [32], [33], [34]

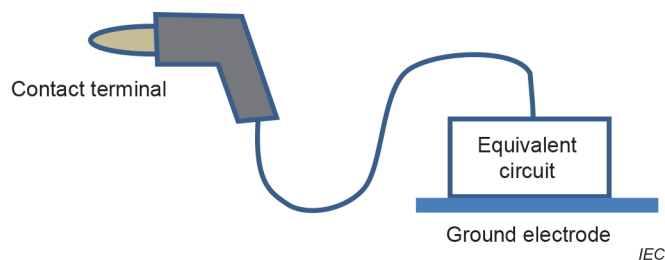


Figure 9 – Example of contact current measurement equipment

6.3.3 Measurements

The following issues shall be considered for the measurement of contact currents:

- Avoid electromagnetic field disturbances caused by the operator and measurement systems as much as possible.
- Environmental conditions (temperature, humidity, vibration, electromagnetic fields, etc.) can affect the measurement equipment.
- The requirements on the measurement equipment shall be specified or the reference to an appropriate document, e.g. IEC TR 63167:2018 [22], shall be included.
- A calibrated measurement system shall be employed.

There are two types of measurements, specified in the following a) and b), depending on the expected exposure scenario.

The measurement equipment specification, exposure scenario and its contact current shall be reported.

- a) Measurement methods using an ungrounded metal board (simulating the EV body):
 - 1) Place a ground electrode where a person can stand.
 - 2) Place an ungrounded metal board large enough at the location where metal objects can exist in the actual usage condition. For representing an EV, a metal board larger than 1,2 m × 1,2 m may be used (see Figure I.10). Choose the closest position in the actual usage because the contact current tends to become higher as metal objects come closer to the DUT. Find the maximum condition of the contact current because the metal board position (height) and direction can change the contact current strength.
 - 3) Touch the contact terminal to the metal board. Find the maximum reading on the meter over multiple touch locations if there is a dependency on the touch position.
- b) Measurement methods using a grounded metal board:
 - 1) Place a grounded metal board large enough to cover all metal objects that may exist in the actual usage condition. Choose the closest position in the actual usage because the contact current tends to become higher as metal objects come closer to the DUT. Find the maximum position of the contact current because the metal board position (height) and direction can change the contact current strength.
 - 2) Touch the contact terminal to the metal board where a person can touch (see Figure I.9). The contact current becomes larger according to the magnitude of the interlinkage magnetic flux to a loop created by the grounded metal board and human body. A rectangular loop with size of minimum 1,5 m (height) × 0,5 m (width) shall be used to evaluate the contact current. Find the loop arrangement which gives the maximum contact current value.

Additional information on the measurement of contact currents can be found in IEC 60990:2016 [30].

7 Computational assessment methods

7.1 General

The internal electric field pE_{ind} , the psSAR, or whole-body SAR for comparison with basic restrictions is calculated by simulation using a representative set of anatomical human models.

NOTE 1 As an example for a representative set, twelve models were used in [35] to derive general coverage factors.

Calculations may be performed using full wave methods or quasi-static methods, depending on the electrical size of the problem geometry. For quasi-static problems, the effect of the induced fields in the human body on the incident field source is assumed to be negligible. The incident fields and the induced fields can be evaluated in separate simulations using different computational methods. For the calculation of the incident fields, analytical or quasi-analytical methods can be applicable. Table 4 summarizes the computational methods for exposure evaluation. A brief description of computational methods is also given in Annex F. Annex H specifies averaging algorithms, and code verification and model validation. At present, body phantoms to conservatively measure the exposure to WPT systems are not available. These models shall include the following:

- availability of larger bodies which have increased absorption rates and are therefore more susceptible to electrostimulation effects from low frequency magnetic fields;
- availability of smaller bodies (children) which can be closer to the field source;
- individual segmentation of the tissues of the central and peripheral nervous systems for evaluation with the respective basic restrictions;

- posable or movable limbs for the modelling of realistic exposure scenarios.

All computational algorithms shall be verified for technically correct implementation. Guidance is provided in Annex H.

NOTE 2 The measurement standard IEC/IEEE 62209-1528:2020 specifies SAR measurement methods for a frequency range from 4 MHz to 10 GHz. While a frequency range of 30 MHz to 6 GHz is specified for the computational standards IEC/IEEE 62704-1:2017 and IEC/IEEE 62704-4:2020, the computational methods are generally independent of the frequency. They are applicable for the frequency range of this document provided that the uncertainty of the truncation of the computational domain is appropriately considered and its uncertainty assessed.

Table 4 – Computational methods

Calculation method	Note
BEM (boundary element method)	Full-wave computation
FDFD (finite difference frequency domain)	Full-wave computation
FDTD (finite difference time domain)	Full-wave computation
FEM (finite element method)	Full-wave or quasi-static approximation computation
QS-FDTD (quasi-static finite difference time domain)	Quasi-static approximation computation
FIT (finite integration technique)	Full-wave or quasi-static approximation computation
MoM (method of moments)	Full-wave computation
Hybrid method (e.g. MoM + FDTD)	Full-wave computation
SPFD (scalar potential finite difference)	Quasi-static approximation computation
IM (impedance method)	Quasi-static approximation computation

7.2 Quasi-static approximation

For the applicability of a computational method based on the quasi-static approximation, the upper frequency limit is established considering the dimensions of the exposed body region or person, the size of the field source, and the distance between them. The frequency limit is then given by the quasi-static approximation of Formula (22), Formula (23), and Formula (24):

$$\left| \omega^2 \varepsilon \mu D^2 \right| \ll 1 \Leftrightarrow \left(\frac{D}{\lambda} \right)^2 \ll 1 \quad (22)$$

or

$$\omega^2 \Re\{\varepsilon\} \mu D^2 \ll 1 \quad (23)$$

and

$$\omega \sigma \mu D^2 \ll 1 \quad (24)$$

where

- ω is the angular frequency;
- $\Re\{\varepsilon\}$ is the real part of the complex relative permittivity of the exposed tissue or the tissue-equivalent medium;
- μ is the permeability;
- λ is the minimum wavelength within the exposed region;
- σ is the conductivity of the exposed tissue or the tissue-equivalent medium;
- D is the feature size, which encompasses the dimensions of the field source and the exposed body or phantom considering the reduction of the wavelength due to the presence of dielectrics with a relative permittivity greater than 1.

Formula (22), Formula (23), and Formula (24) contain only the frequency, the dielectric and magnetic properties, and spatial dimensions of the geometry to be analysed (field source and exposed body). They apply to both electric and magnetic field exposure. The frequency range within which electric and magnetic fields can be regarded as decoupled is identical for electric and magnetic field exposure. Hence, the quasi-static approximation applies to both inductive and capacitive coupling.

A rationale shall be given along with evaluation results on why the quasi-static approximation applies on the evaluated exposure scenario. Possible deviations from a full wave solution shall be discussed and should be considered in the uncertainty budget (9.3).

7.3 Computational assessment against the basic restrictions

7.3.1 General

The basic restrictions on human exposure are specified in terms of:

- peak spatial average SAR [3], [8] (7.3.2);
- whole-body average SAR [3], [8] (7.3.3);
- averaged current density on a surface [4] (Clause G.1);
- peak spatial average E-field in a cubical volume [5] (Clause G.2);
- peak spatial average E-field along a line [8] (G.2.3);
- maximum local E-field [36] (G.2.4).

Different limits can apply for these quantities depending on the exposed body region. The computational code or the applied post-processing techniques shall implement the required averaging methods. Because the descriptions of the averaging algorithms in the exposure guidelines that establish them are generally not sufficiently detailed for an implementation as a computational algorithm, this document refers to the respective computational International Standards that specify those or specifies its own algorithms if an appropriate definition is not available elsewhere. Details are given in 7.3.2, 7.3.3, and Annex G. The implementation of these algorithms shall be validated according to the methods specified in Annex H.

Peak quantities, such as the pE_{ind} or the $psSAR$, can be read out directly from the post processor. Alternatively, the methods specified for the peak spatial average quantities may be applied (7.3.2, Clause G.2) with averaging volumes on the order of magnitude of the voxels or tetrahedra of the exposed body region. Care should be taken concerning stair-casing artefacts because these may affect the readout of the global field maximum or be dominant in small averaging volumes.

7.3.2 Peak spatial-average SAR

The peak spatial average SAR shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the psSAR is evaluated.

7.3.3 Whole-body average SAR

The whole-body average SAR shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the whole-body SAR is evaluated.

7.3.4 Averaged current density on a surface

The electric field distribution in the exposed body shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes. The averaged current density shall then be calculated according to Clause G.1.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the electric field is evaluated.

7.3.5 Peak spatial average internal E-field in a cubical volume

The electric field distribution in the exposed body shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes. The peak spatial average E-field in a cubical volume shall then be calculated according to Clause G.2.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the electric field is evaluated.

7.3.6 Peak spatial average internal E-field along a line

The electric field distribution in the exposed body shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes. The peak spatial average E-field along a line shall then be calculated according to G.2.3.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the electric field is evaluated.

7.3.7 Maximum local internal E-field

The electric field distribution in the exposed body shall be evaluated according to IEC/IEEE 62704-1 for Cartesian meshes, and according to IEC/IEEE 62704-4 for unstructured meshes. The maximum local E-field shall then be calculated according to G.2.4. It should be noted that several simulations with different mesh resolutions according to IEC/IEEE 62704-1 or IEC/IEEE 62704-4 should be used to reduce the impact of staircasing errors.

NOTE The computational methods specified in IEC/IEEE 62704-1 and IEC/IEEE 62704-4 are independent of the frequency at which the electric field is evaluated.

8 Combination of measurement and computational methods for inductive WPT systems

8.1 General

Because internal electric field, current density, and SAR are difficult to measure directly, the evaluation is often performed using computational approaches. However, it is difficult to accurately model an actual exposure situation in simulations including an electric vehicle body, ground planes, etc. In addition, it is sometimes difficult to retrieve information about the size, shape, and design of the WPT systems. It is possible to evaluate internal electric field or SAR by an approach combining both the measurement of electromagnetic fields around the system and computational analyses of induced quantities. The magnetic near-field data are first measured in the area occupied by the human body. Then, the measured magnetic fields are used as incident fields in an applicable computational method (Table 4) to derive the internal electric field strength inside the human-body model.

8.2 Measurement of magnetic field

For measurements of magnetic field distributions below 30 MHz, search-coil sensors are often used. Both amplitude and phase of the magnetic field shall be measured in the area occupied by the human body. To measure the phase of magnetic field, one magnetic sensor is fixed at a specific position to retrieve the reference phase, and the other sensor is moved over the area occupied by the human body.

8.3 Computational analyses of induced quantities

Anatomical human models or appropriate dosimetric phantoms shall be used to calculate induced field quantities. Computational methods for the calculation of the induced field quantities are summarized in Annex F. The measured magnetic field distribution is applied as the incident field in an applicable computational method (Table 4). The measured magnetic field data may be directly applied as a source with some computational methods. However, those computational methods that use vector potentials as a source to calculate the internal electric field require the reconstruction of vector potentials from the measured magnetic field ($\mathbf{H}$) or the magnetic flux density ($\mathbf{B}$) [37]. A finite number of samples of $\mathbf{B}$ from measurements may be used for field reconstruction taken at monitor points. In this case the spatial coordinate of the collected measurement samples is also needed.

The procedure to reconstruct the B-field distribution is specified in [37]. The magnetic flux density is assumed to be $\mathbf{B} = u_x B_x + u_y B_y + u_z B_z$, which satisfies $\nabla \cdot \mathbf{B} = 0$, where $u_{x,y,z}$ are unit vectors. Hence it can be represented using a vector potential $\mathbf{A}$, so that $\nabla \times \mathbf{A} = \mathbf{B}$. A vector potential is given by Formula (25), Formula (26), and Formula (27):

$$A_x = -\int_0^y \left[\frac{1}{3} B_z(x, y', z) + \frac{1}{6} B_z(x, y', 0) \right] dy' + \int_0^z \left[\frac{1}{3} B_y(x, y, z') + \frac{1}{6} B_y(x, 0, z') \right] dz' \quad (25)$$

$$A_y = -\int_0^z \left[\frac{1}{3} B_x(x, y, z') + \frac{1}{6} B_x(0, y, z') \right] dz' + \int_0^x \left[\frac{1}{3} B_z(x', y, z) + \frac{1}{6} B_z(x', y, 0) \right] dx' \quad (26)$$

$$A_z = -\int_0^x \left[\frac{1}{3} B_y(x', y, z) + \frac{1}{6} B_y(x', 0, z) \right] dx' + \int_0^y \left[\frac{1}{3} B_x(x, y', z) + \frac{1}{6} B_x(0, y', z) \right] dy' \quad (27)$$

Formula (25), Formula (26), and Formula (27) require that the vector components of the magnetic flux density be known. Therefore, they are valid for any arbitrary inhomogeneous, anisotropic, or nonlinear medium. In free space, the vector components of the magnetic flux density can be calculated directly from the H-field vector.

8.4 Computational assessment against the basic restrictions

The basic restrictions are generally specified for induced field quantities that are averaged over a certain geometrical shape in the exposed body or phantom for the computational assessment of the induced fields, e.g. pE_{ind} and $psSAR$. The methods described in 8.3 shall be applied to calculate the induced fields for a representative set of anatomical models of the human body or an appropriate dosimetric phantom. Computation of the induced fields against the basic restrictions shall then be carried out according to 7.3.

NOTE Some national regulations specify peak values instead of averaged quantities as basic restrictions on the induced fields.

9 Uncertainty assessments

9.1 General

In 9.2 to 9.4, the sources of measurement and computational uncertainty are listed. Their contributions to the uncertainty shall be assessed for the evaluation of the total uncertainty. The uncertainty budget is based on the concepts of ISO/IEC Guide 98-1 [38], ISO/IEC Guide 98-3 [39], [40], and [41]. For most methods specified in this document, each source of uncertainty is considered as independent and can therefore be assessed separately from the others.

9.2 Measurement methods

9.2.1 Measurement uncertainty budget

Table 5 summarizes the main parameters related to magnetic field measurements.

Table 5 – Example of uncertainty evaluation of the the E-field and H-field exposure assessment using measurement methods

Quantity x_i (source of uncertainty)	Reference	Prob. dist. ^a	Unc.	Divisor ^a q	Standard uncertainty $u(x_i)$	Sensitivity c_i	Expanded uncertainty $u_i(y)$	Degrees of freedom v_i
Amplitude calibration uncertainty	9.2.2	N		1		1		
Probe anisotropy	9.2.3	R		$\sqrt{3}$		1		
Probe dynamic range linearity	9.2.4	R		$\sqrt{3}$		1		
Probe frequency domain response	9.2.5	R		$\sqrt{3}$		1		
Modulation response	9.2.6	R		$\sqrt{3}$		1		
Spatial averaging (maximum gradient)	9.2.7	R		$\sqrt{3}$		1		

Quantity x_i (source of uncertainty)	Reference	Prob. dist. ^a	Unc.	Divisor ^a q	Standard uncertainty $u(x_i)$	Sensitivity c_i	Expanded uncertainty $u_i(y)$	Degrees of freedom v_i
Gradient assessment uncertainty	9.2.8	R		$\sqrt{3}$		1		
Parasitic E-field or H-field sensitivity	9.2.9	R		$\sqrt{3}$		1		
Detection limit	9.2.10	R		$\sqrt{3}$		1		
Readout electronics	9.2.11	N		1		1		
Response time	9.2.12	N		1		1		
Probe positioning	9.2.13	N		1		1		
Signal postprocessing	9.2.14	N		1		1		
Nominal position	9.2.15	R		$\sqrt{3}$		1		
Repeatability	9.2.16	N		1		1		
DUT	9.2.17	N		1		1		
Combined uncertainty ($k = 1$)		RSS						
Expanded uncertainty ($k = 2$) and effective degrees of freedom							$U =$	$v_{\text{eff}} =$
^a Other probability distributions and divisors may be used if they better represent available knowledge of the quantities concerned.								
NOTE 1 Abbreviations: N – normal probability distribution; R – rectangular probability function								
NOTE 2 The divisor is a function of the probability distribution and degrees of freedom (v_i and v_{eff}).								
NOTE 3 c_i is the sensitivity coefficient that is applied to convert the variability of the uncertainty component into a variability of the exposure quantity.								

9.2.2 Amplitude calibration uncertainty

The amplitude calibration uncertainty shall be evaluated within the scope of calibration (frequency and dynamic ranges) of the calibration system, e.g. a transverse electromagnetic (TEM) cell. Details on field generation are given in B.2.1.

9.2.3 Probe anisotropy

The probe anisotropy shall be evaluated by exposure in a homogeneous magnetic field, e.g. inside a TEM cell or appropriate coil configurations. For the quantification of the uncertainty, the probe shall be tilted through angular increments of 5° along the entire specified angular range, and the maximum error shall be reported as uncertainty according to the definition in B.2.2.1.4.

9.2.4 Probe dynamic range linearity

Probe dynamic range shall be evaluated covering the dynamic range from 10 % (–20 dB in terms of induced E-fields or current density, or –10 dB in terms of SAR) the exposure limits of the applicable requirement to greater than four times the occupational exposure limits of the applicable requirement. The probe linearity shall be evaluated in a homogeneous magnetic field, which can be generated in a TEM cell or by appropriate coil configurations. See also B.2.2.1.3.

9.2.5 Probe frequency domain response

The probe frequency domain response is evaluated for a homogenous incident field over specified operational frequency range. The uncertainty due to the probe frequency response is assessed by evaluating the flatness of the probe response across the frequency range. The frequency domain response shall be evaluated in a homogeneous magnetic field, which can be generated in a TEM cell or by appropriate coil configurations. See also B.2.2.1.2.

9.2.6 Modulation response

Probe modulation response shall be evaluated for modulated carriers in the specified frequency range with rectangular pulse modulation, 10 % duty cycle, 10 Hz, 100 Hz, 1 000 Hz pulse repetition rate (whatever is applicable at the carrier frequency). The modulation response shall be evaluated in a homogeneous magnetic field, which can be generated in a TEM cell or by appropriate coil configurations.

9.2.7 Spatial averaging (maximum gradient)

The uncertainty due to spatial averaging shall be evaluated in gradient fields that correspond to those of the incident field of the DUT with respect to amplitude and orientation within a maximum range from 0 T/m/T to at least 30 T/m/T. The gradient fields shall be generated using an appropriate coil configuration (B.2.1) considering the overall dimensions of the probe.

9.2.8 Gradient assessment uncertainty

The uncertainty in the accuracy of field gradient measurements shall be assessed by applying the probe positioning error (9.2.13) on the gradient uncertainty (Clause B.3) and considered as a separate uncertainty parameter. The gradient assessment uncertainty shall only be considered for Tier 3 evaluations.

9.2.9 Parasitic E-field and H-field sensitivity

Parasitic sensitivity is tested with the field probe in maximum hold swept through all possible spherical incident orientations of the field, e.g. in solid angle segments of 30° in ϕ - and θ -directions, with the E-field or H-field amplitude at the occupational exposure limit. Appropriate homogeneous E-fields and H-fields can be generated in a TEM cell or in a parallel plate capacitor. Details on field generation are given in B.2.1.

9.2.10 Detection limit

Detection limit is the contribution of the readout electronic noise, i.e. the signal-to-noise ratio, at 10 % (–20 dB in terms of induced E-fields or current density, or –10 dB in terms of SAR) of the general public exposure limits of the applicable requirement.

9.2.11 Readout electronics

Readout electronics is the uncertainty contribution by analogue-to-digital conversion.

9.2.12 Response time

If the probe is operated in swept mode, response time is the sampling response error for the probe swept through a gradient field of at least 20 T/m/T at a speed of 0,1 m/s.

9.2.13 Probe positioning

The uncertainty due to probe positioning shall be evaluated as the error in a gradient field of at least 30 T/m/T considering the mechanical tolerances, such as the positioning accuracy, of the probe positioning system (e.g. robot).

9.2.14 Signal postprocessing

The uncertainty of postprocessing (shaping, filtering, and signal conditioning) has to be assessed using theoretical or experimental evaluation methods.

9.2.15 Nominal position

The nominal positioning uncertainty shall be evaluated as the positioning uncertainty (9.2.13) with respect to assessing the worst-case exposure.

9.2.16 Repeatability

The repeatability shall be evaluated by measuring the field amplitude at a specific point at the closest accessible distance to the source ten times. It is then evaluated as the standard deviation of these measurements.

9.2.17 DUT

The uncertainty associated with the DUT shall be carefully evaluated and assessed. These include drift of amplitude, spectrum, intermittency of emission. In addition to the uncertainty of the steady-state exposure relevant for SAR, the uncertainty of the peak transient field assessment also has to be assessed.

9.3 Computational methods

9.3.1 Computational uncertainty budget

Table 6 summarizes the sources of uncertainty for computational methods. The methodology closely follows IEC/IEEE 62704-1 and IEC/IEEE 62704-4. These documents are referenced in 9.3 where applicable.

NOTE While the computational standards IEC/IEEE 62704-1, IEC/IEEE 62704-4, and IEC/IEEE 63195-2 [42] are developed for particular applications and frequency ranges, the computational techniques and the procedures for their verification and uncertainty assessment are generally independent of the frequency ranges for which the respective standards are specified. They can therefore be applied without modification. Particular distances or dimensions are scaled in accordance with the frequency range of this document.

Table 6 – Example of uncertainty evaluation of computational methods

Quantity x_i (source of uncertainty)	Reference	Prob. dist. ^a	Unc.	Divisor ^a q	Standard uncertainty $u(x_i)$	Sensitivity c_i	Expanded uncertainty $u_i(y)$	Degrees of freedom ν_i
Numerical uncertainties								
Grid resolution	9.3.2	R		$\sqrt{3}$		1		
Tissue parameters	9.3.3	R		$\sqrt{3}$		1		
Exposure position	9.3.4	R		$\sqrt{3}$		1		
Convergence	9.3.5	R		$\sqrt{3}$		1		
Power budget	9.3.6							
Boundary conditions	9.3.7	R		$\sqrt{3}$		1		
Quasi-static approximation	9.3.8	R		$\sqrt{3}$		1		
Modelling uncertainties								
Model parts and geometry	9.3.9							
Dielectric parameters	9.3.10							
Ferrite parameters	9.3.11							
Positioning of transmit and receive coils	9.3.12							
Coupling of transmit and receive coils	9.3.13							
Exposure sources other than the coils	9.3.14							
Loading of the coil	9.3.15							
Combined uncertainty ($k = 1$)		Root sum square (RSS)						
Expanded uncertainty ($k = 2$) and effective degrees of freedom							$U =$	$\nu_{\text{eff}} =$
^a Other probability distributions and divisors may be used if they better represent available knowledge of the quantities concerned.								
NOTE 1 Abbreviations: N – normal probability distribution; R – rectangular probability function								
NOTE 2 The divisor is a function of the probability distribution and degrees of freedom (ν_i and ν_{eff}).								
NOTE 3 c_i is the sensitivity coefficient that is applied to convert the variability of the uncertainty component into a variability of the exposure quantity.								

9.3.2 Grid resolution

For rectilinear grids, the impact of the resolution shall be determined by reducing the step sizes of all axes by factors of 2 then 4. If the simultaneous reduction of the step sizes is not possible, for example, because of memory limitations, x -, y - and z -axes shall be refined separately by factors of 2 then 4.

For the coarsest grid resolution, at least two mesh cells shall be placed between the coil windings.

For tetrahedral meshes, the impact of the resolution shall be determined in accordance with one of the following methods.

- a) If the applied computational code does not employ adaptive mesh refinement, the total number of mesh elements used for the initial mesh shall be increased iteratively by at least 50 %. The tetrahedral shall be refined equivalently over the entire computational domain.
- b) If the applied computational code uses adaptive mesh refinement, iterative adaptive passes shall be performed until the total number of elements has increased by at least 20 % for second or third lowest order and 50 % for lowest order.

The deviation of the target quantity for all test cases shall be reported, and the uncertainty shall be determined from the maximum deviation of all simulated cases. If the mesh axes of rectilinear grids are refined separately, the root sum square (RSS) of the maxima of each axis shall be reported with rectangular probability distribution.

9.3.3 Tissue parameters

The tissue dielectric properties used for anatomical models are generally reported with an uncertainty value (e.g. [43], [44]). The impact of these uncertainties shall be evaluated by applying the minimum and maximum conductivity and permittivity (four different combinations) using the reported uncertainty limits in comparison with the results obtained for the nominal dielectric property values. The maximum uncertainty shall be reported. When modelling dosimetric phantoms with reference dielectric properties, such as those specified for tissue-equivalent medium, the uncertainty of the phantom medium dielectric properties shall be assumed as zero. The maximum deviation of the target quantity shall be reported with rectangular probability distribution.

NOTE The permittivity of the tissue or tissue-equivalent medium is always assumed to be real, i.e. the imaginary part of the complex relative permittivity always refers to the conductivity only.

9.3.4 Exposure position

For Cartesian meshes, the distance between the phantom the body model or the location of the evaluation points and the DUT shall be varied by ± 1 mesh step sizes. In meshes with varying mesh step size, the shifting distance shall be chosen in accordance with the mesh step at the side of the bounding box of the phantom, the body model or the evaluation points that faces the source. The maximum deviation shall be reported with rectangular probability distribution.

For tetrahedral meshes, no positioning uncertainty applies because these meshes precisely adapt to the geometry of the DUT and the phantom.

9.3.5 Convergence

For time domain simulations, the convergence uncertainty shall be evaluated according to the methods of IEC/IEEE 62704-1. For frequency domain simulations, the convergence uncertainty shall be evaluated according to the methods of IEC/IEEE 62704-4. If a full-wave method is applied, the power budget shall be evaluated according to IEC/IEEE 62704-1, and its contribution shall be added to the convergence uncertainty.